

More on FT

Lecture 10

4CT.5

3CT.3-5,7,8

Higher Order Differentiation

$$\mathfrak{T}\left[\frac{d^n f(t)}{dt^n}\right] = (j\omega)^n F(j\omega)$$

$$\sum_{n=0}^N a_n \frac{d^n y(t)}{dt^n} = \sum_{m=0}^M b_m \frac{d^m x(t)}{dt^m}, \Rightarrow \mathfrak{T}\left\{\sum_{n=0}^N a_n \frac{d^n y(t)}{dt^n}\right\} = \mathfrak{T}\left\{\sum_{m=0}^M b_m \frac{d^m x(t)}{dt^m}\right\}$$

$$\sum_{n=0}^N a_n (j\omega)^n Y(j\omega) = \sum_{m=0}^M b_m (j\omega)^m X(j\omega)$$

$$Y(j\omega) = \frac{\sum_{m=0}^M b_m (j\omega)^m}{\sum_{n=0}^N a_n (j\omega)^n} X(j\omega)$$

$$(a_n p^n + a_{n-1} p^{n-1} + \dots + a_1 p^1 + a_0) y(t) = (b_m p^m + \dots + a_0) x(t)$$

$$y(t) = \frac{B(p)}{A(p)} x(t)$$

$$Y(j\omega) = \frac{\mathbf{B}(j\omega)}{\mathbf{A}(j\omega)} \mathbf{X}(j\omega) = \mathbf{H}(j\omega) \mathbf{X}(j\omega)$$

Frequency Response

- We just have seen that the FT of the differential equation describing a system yields the Frequency Response, $\mathbf{H}(j\omega)$
- We have also seen that convolving the impulse response, $h(t)$, with any input, $x(t)$ will yield its output, $y(t)$.
- Is the Frequency Response related to the impulse response?

Frequency Response and the Impulse Response

Using convolution, for only sinusoidal inputs:

$$x(t) = Ae^{j(\omega t + \phi)}$$

$$y(t) = \int h(\tau)x(t - \tau)d\tau = \int h(\tau)Ae^{j[\omega(t - \tau) + \phi]}d\tau$$

$$= \int h(\tau)Ae^{j(\omega t + \phi)}e^{-j\omega\tau}d\tau = Ae^{j(\omega t + \phi)} \int h(\tau)e^{-j\omega\tau}d\tau = x(t) \int h(\tau)e^{-j\omega\tau}d\tau$$

$$y(t) = x(t) \int h(\tau)e^{-j\omega\tau}d\tau$$

Taking the FT of both sides and noting that $\int h(\tau)e^{-j\omega\tau}d\tau$ is not a function of t :

$$\int y(t)e^{-j\omega t}dt = \int x(t)e^{-j\omega t}dt \int h(\tau)e^{-j\omega\tau}d\tau$$

$$Y(j\omega) = X(j\omega) \int h(\tau)e^{-j\omega\tau}d\tau \Rightarrow H(j\omega) = \frac{Y(j\omega)}{X(j\omega)} = \int h(\tau)e^{-j\omega\tau}d\tau$$

But $\int h(\tau)e^{-j\omega\tau}d\tau$ is the FT of $h(t)$ and, therefore,

the frequency response is just the FT of the impulse response.

We say that $H(j\omega)$ and $h(t)$ are FT pairs.

Another Interesting Example

$$\begin{aligned}f(t) &= V \text{ for } -\tau/2 < t < \tau/2; \\&= V[u(t + \tau/2) - u(t - \tau/2)] \\F(j\omega) &= \int_{-\infty}^{\infty} f(t)e^{-j\omega t} dt \\&= \int_{-\tau/2}^{\tau/2} Ve^{-j\omega t} dt = \frac{V}{-j\omega} e^{-j\omega t} \Big|_{-\tau/2}^{\tau/2} = \frac{V}{-j\omega} (e^{-j\omega\tau/2} - e^{j\omega\tau/2}) \\&= \frac{V}{j\omega} (e^{j\omega\tau/2} - e^{-j\omega\tau/2}) = 2\frac{V}{\omega} \left(\frac{e^{j\omega\tau/2} - e^{-j\omega\tau/2}}{2j} \right) = 2\frac{V}{\omega} \sin(\omega\tau/2) \\&= V\tau \frac{\sin(\omega\tau/2)}{\omega\tau/2} = V\tau \text{ Sa}(\omega\tau/2)\end{aligned}$$

- FT of rectangular pulse is the Sampling Function
- What is the FT of $u(t)$?

FT of the Unit Impulse

$$\mathfrak{T}[\delta(t)] = \int_{-\infty}^{\infty} \delta(t) e^{-j\omega t} dt$$

$$\text{From } f(x) = \int_{-\infty}^{\infty} f(t) \delta(t-x) dt$$

$$\mathfrak{T}[\delta(t)] = e^{-j\omega 0} = 1$$

- If $x(t) = \delta(t)$, then $y(t) = h(t)$ the response due to a unit impulse response and $H(j\omega)$ is the network response or system function in phasor form

$$Y(j\omega) = \frac{B(j\omega)}{A(j\omega)} X(j\omega); \quad X(j\omega) = 1; \quad \therefore \frac{B(j\omega)}{A(j\omega)} = H(j\omega)$$

This implies that the FT of the impulse response is the network response or frequency response

$$H(j\omega) = \mathfrak{T}[h(t)] = \int_{-\infty}^{\infty} h(t) e^{-j\omega t} dt$$

FT of $\delta(t)$ continued

We just saw that $\mathfrak{F}[\delta(t)] = e^{-j\omega 0} = 1$

therefore, the inverse Fourier transform of 1 must be the delta function.

$$\mathfrak{F}^{-1}[1] = \delta(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} 1e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\infty}^{\infty} (\cos \omega t + j \sin \omega t) d\omega$$
 Using the definition

of the inverse FT.

$$= \frac{1}{2\pi} \int_{-\infty}^{\infty} \cos \omega t d\omega + \frac{1}{2\pi} \int_{-\infty}^{\infty} j \sin \omega t d\omega$$

$$= \frac{1}{2\pi} \int_{-\infty}^{\infty} \cos \omega t d\omega = \frac{1}{\pi} \int_0^{\infty} \cos \omega t d\omega; \text{ using the properties of even and odd functions.}$$

$$\delta(t) = \frac{1}{\pi} \int_0^{\infty} \cos \omega t d\omega$$

Odd and Even Properties of a Signal

Note that the sine is odd and $\lim_{a \rightarrow \infty} \int_{-a}^a \sin \omega t d\omega = \lim_{a \rightarrow \infty} \left\{ \int_0^a \sin \omega t d\omega + \int_{-a}^0 \sin \omega t d\omega \right\}$

If $\int_0^a \sin \omega t d\omega = g(a)$; then $\int_{-a}^0 \sin \omega t d\omega = -g(a)$ since

Let $\omega = -x, \Rightarrow \int_a^0 \sin(-xt)(-dx) = \int_a^0 -\sin(xt)(-dx) = \int_a^0 \sin(xt)dx = -\int_0^a \sin(xt)dx = -g(a)$

$$\lim_{a \rightarrow \infty} \{g(a) - g(a)\} = 0$$

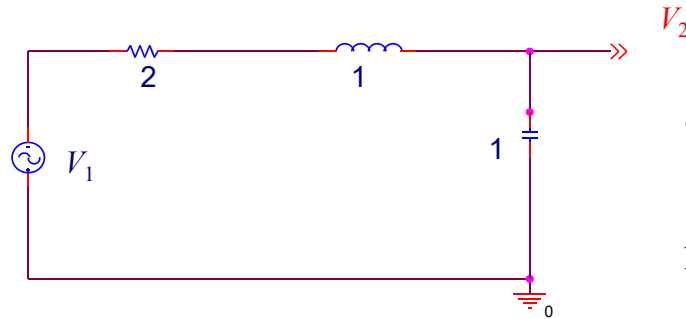
Note that the cosine is even and $\lim_{a \rightarrow \infty} \int_{-a}^a \cos \omega t d\omega = \lim_{a \rightarrow \infty} \left\{ \int_0^a \cos \omega t d\omega + \int_{-a}^0 \cos \omega t d\omega \right\}$

If $\int_0^a \cos \omega t d\omega = g(a)$; then $\int_{-a}^0 \cos \omega t d\omega = g(a)$ since

Let $\omega = -x, \Rightarrow \int_a^0 \cos(-xt)(-dx) = \int_a^0 \cos(xt)(-dx) = -\int_a^0 \cos(xt)dx = \int_0^a \cos(xt)dx = g(a)$

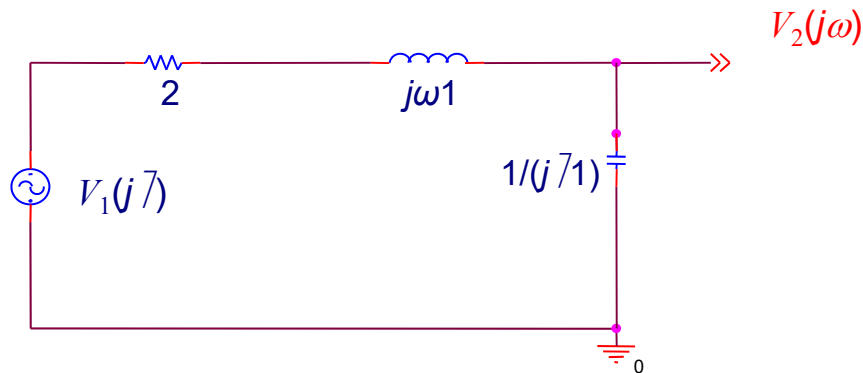
$$\lim_{a \rightarrow \infty} \{g(a) + g(a)\} = \lim_{a \rightarrow \infty} \{2g(a)\} = \lim_{a \rightarrow \infty} 2 \int_0^a \cos \omega t d\omega$$

Calculation of the Frequency Response



Given that $V_1(t) = \delta(t)$ and it is known that $V_2(t) = te^{-t} u(t)$ what is FT of $[te^{-t} u(t)]$ which must equal be $H(j\omega)$?

We can calculate FT of $[te^{-t} u(t)]$ by evaluating the integral of $te^{-t} u(t) e^{-j\omega t}$ or calculate the system response function:



$$V_2(j\omega) = \frac{1/j\omega}{2 + j\omega + 1/j\omega} V_1(j\omega)$$

$$H(j\omega) = \frac{1/j\omega}{2 + j\omega + 1/j\omega} = \frac{1}{2j\omega - \omega^2 + 1}$$

$$= \frac{1}{(1 + j\omega)^2} = \mathfrak{F}[te^{-t} u(t)]$$

Fourier Transform of A Constant

$$\mathfrak{F}[1] = \int_{-\infty}^{\infty} e^{-j\omega t} dt = \int_{-\infty}^{\infty} (\cos \omega t - j \sin \omega t) dt$$

Since a constant is an even function over ∞ :

$$= \int_{-\infty}^{\infty} \cos \omega t dt$$

$$\text{But we said that } \delta(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} \cos \omega t d\omega$$

$$\therefore \mathfrak{F}[1] = 2\pi\delta(\omega)$$

- Our first Duality:
 - A constant in the time domain is a unit impulse function in the frequency domain
 - A unit impulse function in the time domain is a constant in the frequency domain

FT of a Unit Step Function

$$\begin{aligned}\mathfrak{F}[u(t)] &= \int_{-\infty}^{\infty} u(t)e^{-j\omega t} dt = \int_0^{\infty} e^{-j\omega t} dt = \frac{e^{-j\omega t}}{-j\omega} \Big|_0^{\infty} \\ &= \frac{1}{j\omega} (1 - e^{-j\omega\infty}), \text{ which is undefined}\end{aligned}$$

Let's use another technique :

$$\mathfrak{F}\left[\frac{df(t)}{dt}\right] = j\omega F(j\omega)$$

$$\mathfrak{F}[\delta(t)] = 1 = \mathfrak{F}\left[\frac{du(t)}{dt}\right] = j\omega\mathfrak{F}[u(t)]$$

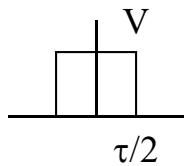
$$\therefore \mathfrak{F}[u(t)] = \frac{1}{j\omega}, \text{ for } \omega \neq 0$$

Time & Frequency Displacement

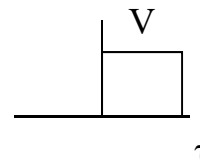
Time Displacement

If $\mathfrak{F}[f(t)] = F(j\omega)$, then what is $\mathfrak{F}[f(t - T_d)]$

$$\begin{aligned}
 &= \int_{-\infty}^{\infty} f(t - T_d) e^{-j\omega t} dt, \text{ let } x = t - T_d \\
 &= \int_{-\infty}^{\infty} f(x) e^{-j\omega(x+T_d)} dx = e^{-j\omega T_d} \int_{-\infty}^{\infty} f(x) e^{-j\omega x} dx \\
 &= F(j\omega) e^{-j\omega T_d}
 \end{aligned}$$



$$\begin{aligned}
 V_1(j\omega) &= V \tau \text{Sa}(\omega\tau/2) \\
 &= V \tau \frac{\sin(\omega\tau/2)}{\omega\tau/2}
 \end{aligned}$$



$$\begin{aligned}
 V_2(j\omega) &= V_1(j\omega) e^{-j\omega\tau/2} \\
 &= V \tau \text{Sa}(\omega\tau/2) e^{-j\omega\tau/2} \\
 &= V \tau \frac{\sin(\omega\tau/2)}{\omega\tau/2} e^{-j\omega\tau/2} \\
 &= V \frac{(1 - e^{-j\omega\tau})}{j\omega}
 \end{aligned}$$

Time & Frequency Displacement

Frequency Displacement

If $\mathfrak{F}[f(t)] = F(j\omega)$, then what is $F(\alpha + j\omega)$

Assume there exist a duality

between Time and Frequency displacement and try

$$\begin{aligned}\mathfrak{F}[f(t)e^{-\alpha t}] &= \int_{-\infty}^{\infty} f(t)e^{-\alpha t} e^{-j\omega t} dt \\ &= \int_{-\infty}^{\infty} f(t)e^{-(\alpha+j\omega)t} dt = F(\alpha + j\omega)\end{aligned}$$

- Our next Duality:
 - A shift in the time domain equals a multiplication by $e^{-j\omega T}$ the frequency domain
 - A shift in the frequency domain equals a multiplication by $e^{-\alpha t}$ the time domain

FT Symmetries

Note: $F(j\omega) = \int_{-\infty}^{\infty} f(t)e^{-j\omega t} dt$

And: $F(-j\omega) = \int_{-\infty}^{\infty} f(t)e^{j\omega t} dt$

And: $F^*(j\omega) = \int_{-\infty}^{\infty} f(t)e^{j\omega t} dt$

Which means: $F^*(j\omega) = F(-j\omega)$

$\therefore \text{Re}[F(j\omega)]$ is even, $\text{Im}[F(j\omega)]$ is odd

Proof:

$$F(j\omega) = A(\omega) + jB(\omega)$$

$$F^*(j\omega) = A(\omega) - jB(\omega)$$

$$F(-j\omega) = A(-\omega) + jB(-\omega)$$

Since $F^*(j\omega) = F(-j\omega)$ then

$$A(\omega) - jB(\omega) = A(-\omega) + jB(-\omega)$$

$$\text{and } A(\omega) = A(-\omega) \text{ \& } -B(\omega) = B(-\omega)$$

Some Interesting FTs

Exponential Functions

$$\mathfrak{F}[e^{-\alpha t} u(t)] = \frac{1}{\alpha + j\omega}, \text{ since } \mathfrak{F}[u(t)] = \frac{1}{j\omega}$$

$$\mathfrak{F}[e^{-\alpha t} tu(t)] = \frac{1}{(\alpha + j\omega)^2}, \text{ since } \mathfrak{F}[tu(t)] = \frac{1}{(j\omega)^2},$$

$$\text{since } \mathfrak{F}\left[\frac{dtu(t)}{dt}\right] = j\omega \mathfrak{F}[tu(t)] = \mathfrak{F}[u(t)] = \frac{1}{j\omega}$$

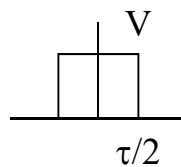
More Interesting FTs

$f(t) = m(t) \cos \omega_0 t$ This is called amplitude modulation

$$= \frac{m(t)}{2} e^{j\omega_0 t} + \frac{m(t)}{2} e^{-j\omega_0 t}$$

$$\mathfrak{F}[f(t)] = \frac{1}{2} M[j(\omega - \omega_0)] + \frac{1}{2} M[j(\omega + \omega_0)]$$

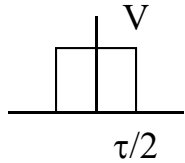
This says that we have made two copies of the spectrum of $m(t)$ one shifted to the right of the $\omega=0$ axis and one to the left. Both of these copies are reduced by $1/2$.



$$\begin{aligned} M(j\omega) &= V\tau \text{Sa}(\omega\tau/2) \\ &= V\tau \frac{\sin(\omega\tau/2)}{\omega\tau/2} \end{aligned}$$

Let's look at the case where $m(t)$ is a pulse and has the spectrum shown above.

Frequency Shifting Example



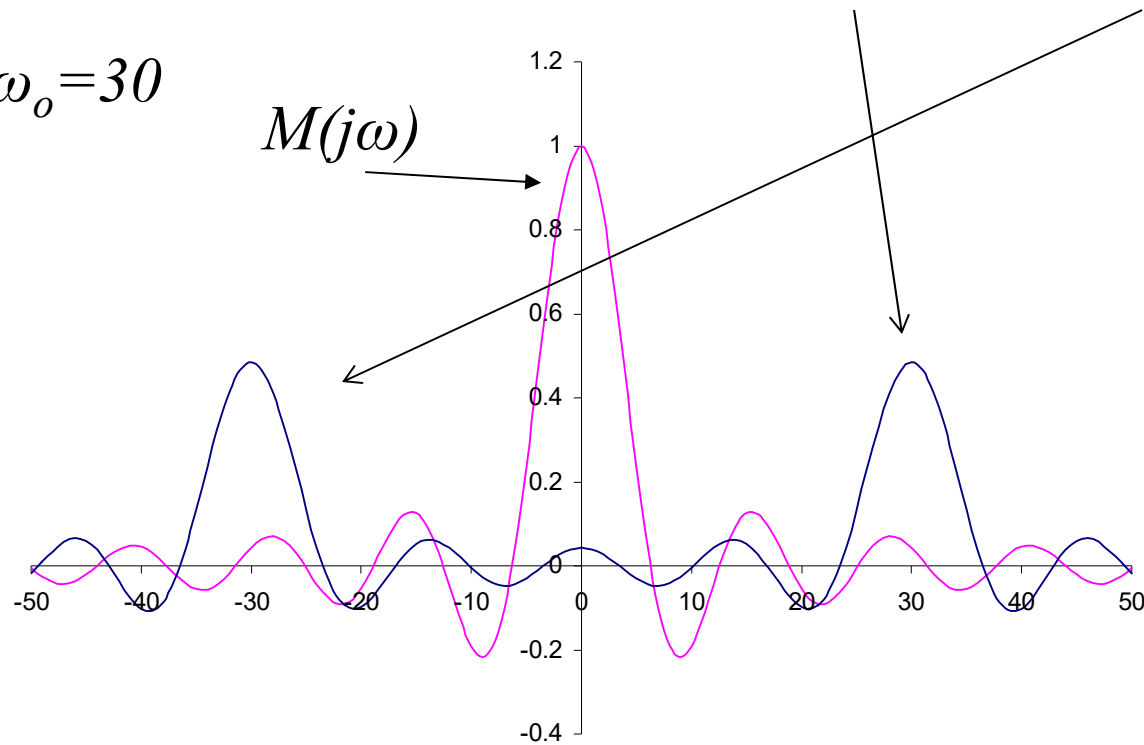
$$M(j\omega) = V\tau \text{Sa}(\omega\tau/2)$$

$$= V\tau \frac{\sin(\omega\tau/2)}{\omega\tau/2}$$

$$F(j\omega) = \frac{1}{2}M[j(\omega - \omega_0)] + \frac{1}{2}M[j(\omega + \omega_0)]$$

$$= \frac{V\tau}{2} \left\{ \frac{\sin[(\omega - \omega_0)\tau/2]}{(\omega - \omega_0)\tau/2} + \frac{\sin[(\omega + \omega_0)\tau/2]}{(\omega + \omega_0)\tau/2} \right\}$$

$$V = \tau = 1, \omega_0 = 30$$



Fourier Transform and Fourier Series

	FT	FS
Time	Aperiodic	Periodic
Frequency	Continuous	Discrete

Now what about FT of a periodic function

Recall:

$$\mathfrak{F}[\delta(t)] = 1$$

$$\mathfrak{F}[1] = 2\pi\delta(\omega)$$

$$\mathfrak{F}[e^{j\omega_0 t}] = \mathfrak{F}[1e^{j\omega_0 t}] = 2\pi\delta(\omega - \omega_0) \text{ (Frequency Shift)}$$

$$\mathfrak{F}[\cos \omega_0 t] = \mathfrak{F}\left[\frac{1}{2}(e^{j\omega_0 t} + e^{-j\omega_0 t})\right] = \pi[\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$$

$$\mathfrak{F}[\sin \omega_0 t] = \mathfrak{F}\left[\frac{1}{2j}(e^{j\omega_0 t} - e^{-j\omega_0 t})\right] = -j\pi[\delta(\omega - \omega_0) - \delta(\omega + \omega_0)]$$

FT of a Periodic Function

FT of a Periodic Function $f(t)$

A periodic function can be formulated in a FS: $f(t) = \sum_{-\infty}^{\infty} a_k e^{jk2\pi t/T}$

$$\therefore \mathfrak{F}[f(t)] = F(j\omega) = 2\pi \sum_{-\infty}^{\infty} a_k \delta(\omega - \frac{2\pi}{T}k)$$

\therefore FT of a periodic function is a series of unit impulse functions and is discrete in the frequency domain.

Note: FT of a train of unit impulses $f(t) = \sum_{-\infty}^{\infty} A_n \delta(t - nT)$

$$\mathfrak{F}[f(t)] = F(j\omega) = \sum_{-\infty}^{\infty} A_n e^{-j\omega nT}$$

\therefore FT of a train of unit impulses is a FS in the frequency domain and is continuous in the frequency domain

Energy of $f(t)$

- Energy dissipated by, say, a 1 ohm resistance over a period T is
$$\int_{-T/2}^{T/2} v(t)^2 dt$$

We also called this the quadratic content of a signal

- It can be shown that the quadratic content of a signal in terms of its FT is:

$$\int_{-T/2}^{T/2} f(t)^2 dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(j\omega)F^*(j\omega)d\omega$$

- And

$$\int_{-T/2}^{T/2} |f(t)|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} |F(j\omega)|^2 d\omega$$

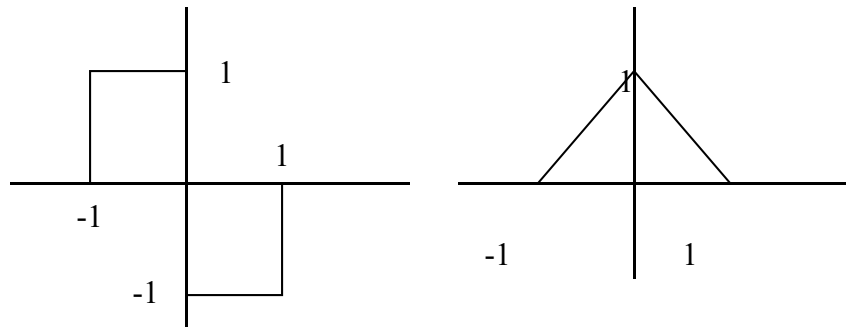
- The latter equation is called Parseval's Theorem

Homework

- Problem (1)
 - Show that $F(j\omega)$ of an even function $f(t)=f(-t)$ is real
 - Show that $F(j\omega)$ of an odd function $f(t)=-f(-t)$ is imaginary
- Problem (2)
 - Find $F(j\omega)$ for the following functions and sketch their amplitude and phase spectra:
 - $f_a(t)=e^{-\alpha t} u(t)$;
 - $f_b(t)= e^{\alpha t} u(-t)+e^{-\alpha t} u(t)$;
 - $f_c(t)= -e^{\alpha t} u(-t)+e^{-\alpha t} u(t)$;
 - $f_d(t)= e^t \sin 10t u(-t)$

Homework

- Problem (3)
 - Calculate the $F(j\omega)$ for the waveforms below. Note that the second is the integral of the first.



- 4CT.5.1, 4CT.5.2
- 3CT.5.1, 3CT.5.2, 3CT.5.3