

# *LTI ODE Continued*

## Lecture #4

# *Introduction of the $p$ Operator*

Let's start with this 2nd order differential equation  
to represent some system.

$$a \frac{d^2 y(t)}{dt^2} + b \frac{dy(t)}{dt} + cy(t) = e \frac{dx(t)}{dt} + fx(t)$$

where  $x(t)$  is the input or source,  
 $y(t)$  is the solution or the response of the differential equation,  
and  $a, b, c, d, f$  are the coefficients.

Let's define the  $p$  operators:

$$p \cong \frac{d}{dt}, p^2 \cong \frac{d^2}{dt^2}, \dots, p^n \cong \frac{d^n}{dt^n}$$

and rewrite the differential equation as

$$ap^2 y(t) + bpy(t) + cy(t) = ep x(t) + fx(t)$$

## *Introduction of the p Operator*

And now for some mathematical blasphemy!!!

$$ap^2 y(t) + bpy(t) + cy(t) = ep x(t) + fx(t)$$

$$[ap^2 + bp + c]y(t) = [ep + f]x(t)$$

$$y(t) = \frac{ep + f}{ap^2 + bp + c} x(t)$$

This is **not** a solution for  $y(t)$  but a way of reducing an equation into a simpler form.

## *Introduction of the $p$ Operator*

Note that  $[ap^2 + bp + c]$  and  $[ep + f]$  are polynomials in  $p$  and we can rewrite these polynomials as  $[ap^2 + bp + c] = A(p)$  and  $[ep + f] = B(p)$  and we get

$$A(p)y(t) = B(p)x(t)$$

$$y(t) = \frac{B(p)}{A(p)}x(t)$$

This is **not** a solution for  $y(t)$  but a way of reducing an equation into a simpler form.

# *Components of the Solution of LODE*

$$A(p)y(t) = B(p)x(t)$$

Let's define  $y(t)$  which is  
the total response of the system

$$y(t) = y_s(t) + y_{sf}(t)$$

where  $y_s(t)$  is called the response due to the source

and is the solution to this

$$A(p)y_s(t) = B(p)x(t)$$

and  $y_{sf}(t)$  is called the source free response or the transient response.

and is the solution to this equation

$$A(p)y_{sf}(t) = 0$$

We can now use superposition, solve two simpler equations,  
and add them up to get the total solution.

## *Components of the Solution of LODE*

### *Source Response*

- $y_s(t)$  - Source Response has the same form as the source and is a solution of:

$$A(p)y_s(t) = B(p)x(t)$$

- If  $x(t)$  is a constant then  $y_s(t)$  is a constant, if  $x(t)$  is a polynomial then  $y_s(t)$  is a polynomial, if  $x(t)$  is a sinusoid then  $y_s(t)$  is a sinusoid, etc.

# *Components of the Solution of LODE*

## *Source-Free Response*

- $y_{sf}(t)$  - Source-free Response is also called the Natural Mode Response, Transient Response, Homogeneous, or Characteristic Response and is a solution of

$$A(p)y_{sf}(t)=0 \text{ Homogeneous Equation}$$

$$A(p)=0 \text{ Characteristic Equation}$$

- Functions which satisfy the Homogeneous Equation are called eigenfunctions (e.g.,  $Ke^{at}$ )
- The values of  $p$  which are solutions to the Characteristic Equation are called eigenvalues.

# *Components of the Solution of LODE*

## *Source-Free Response*

### *Stability*

- The source-free response is independent of the source and always appears. It is a function of the system under examination.
- It determines the stability of the system.
- For a stable system,
  1. it is expected that the source-free response is also known as the transient response since it is expected that this component will effectively “end”.
  2. When the source-free response ends, the response due to the source or steady state remains.
- For an unstable system, the source-free response may never end.



# ***Components of the Solution of LODE: Source Response***

Let us look back at the RL circuit 1<sup>st</sup> Order ODE:  $V_S = i(t)R + L \frac{di(t)}{dt}$

Source Response:  $i_s(t)$

Since the source is a constant,  $V_S$ , then source response is a constant

$$i_s(t) = K$$

$$\frac{di_s(t)}{dt} + \frac{R}{L} i_s(t) = \frac{V_S}{L}$$

Substituting  $i_s(t)$  into the differential equation, we have

$$K = \frac{V_S}{R}$$
$$i_s(t) = \frac{V_S}{R}$$

# ***Components of the Solution of LODE: Source Free Response: Using the Homogenous Equation***

Source Free Response:  $i_{sf}(t)$

The Homogenous equation:  $L\frac{di(t)}{dt} + i(t)R = 0$

Solutions of the Homogenous equation (the eigenfunctions for a 1<sup>st</sup> order ODE) are

$$i_{sf}(t) = Ae^{at}$$

Substituting  $i_{sf}(t)$  into the homogenous equation, we have

$$\frac{di_{sf}(t)}{dt} = aAe^{at}$$

$$aAe^{at} + \frac{R}{L}Ae^{at} = 0$$

$$a = -\frac{R}{L}$$

$$\therefore i_{sf}(t) = Ke^{-\frac{R}{L}t}$$

# ***Components of the Solution of LODE: Source Free Response: Using the Characteristic Equation***

Source Free Response:  $i_{sf}(t)$

Note that the solution of the Characteristic equation (the eigenvalue) is

$$\left(p + \frac{R}{L}\right) = 0$$

$$p = -\frac{R}{L}$$

and, therefore, same solution:  $i_{sf}(t) = Ae^{-\frac{R}{L}t}$

# *Components of the Solution of LODE: Total Solution*

Then the total response is:

$$\begin{aligned}i(t) &= i_s(t) + i_{sf}(t) \\ &= \frac{V_S}{R} + Ae^{-\frac{R}{L}t}\end{aligned}$$

The constant  $A$  can be found from initial conditions of  $i(t)$

$$i(0) = \frac{V_S}{R} + Ae^{-\frac{R}{L}0} = \frac{V_S}{R} + A$$

$$A = i(0) - \frac{V_S}{R}$$

$$i(t) = \frac{V_S}{R} + \left(i(0) - \frac{V_S}{R}\right)e^{-\frac{R}{L}t}$$

## *Components of the Solution of LODE*

In our problem  $i(0) = 0$  :

$$i(t) = \frac{V_S}{R} + \left( i(0) - \frac{V_S}{R} \right) e^{-\frac{R}{L}t}$$

$$i(t) = \frac{V_S}{R} - \frac{V_S}{R} e^{-\frac{R}{L}t}$$

$$= \frac{V_S}{R} \left( 1 - e^{-\frac{R}{L}t} \right)$$

# *Free Response of a 2<sup>nd</sup> ODE*

## *Solutions of the Characteristic Equation*

$$ap^2 + bp + c = 0$$

Let  $p_1, p_2$  be the roots of the characteristic equation, then

$$p_{1,2} = \frac{-b \pm \sqrt{b^2 - 4ac}}{2a}$$

# *The 4 Free Response Cases of a 2<sup>nd</sup> ODE Solutions of the Homogeneous Equation*

Case	Roots		Solution	Type
$b^2 - 4ac > 0$	$p_{1,2} = \frac{-b \pm \sqrt{b^2 - 4ac}}{2a}$	Real, Unequal, Negative	$y_{sf}(t) = C_1 e^{p_1 t} + C_2 e^{p_2 t}$ where $C_1$ and $C_2$ are real	Overdamped Stable
$b^2 - 4ac = 0$	$p_{1,2} = p = \frac{-b}{2a}$	Real, Equal, Negative	$y_{sf}(t) = (C_1 t + C_2) e^{pt}$ where $C_1$ and $C_2$ are real	Critically damped Stable
$b^2 - 4ac < 0$	$p_{1,2} = \frac{-b \pm j\sqrt{4ac - b^2}}{2a}$ $= \alpha \pm j\omega$	Complex conjugates, Unequal	$y_{sf}(t) = e^{-\alpha t} (C_1 e^{j\omega t} + C_2 e^{-j\omega t})$ where $p_{1,2} = -\alpha \pm j\omega$ $C_2 = C_1^*$ ; $C_1 = C e^{j\theta}$ are complex conjugates $y_{sf}(t) = e^{-\alpha t} C (e^{j(\omega t + \theta)} + e^{-j(\omega t + \theta)})$ $= e^{-\alpha t} 2C \cos(\omega t + \theta)$	Underdamped Stable
$b^2 - 4ac < 0$ & $b = 0$	$p_{1,2} = \pm j \frac{\sqrt{4ac}}{2a} = \pm j\omega$	Imaginary	$y_{sf}(t) = C_1 e^{j\omega t} + C_2 e^{-j\omega t}$ $= 2C \cos(\omega t + \theta)$ where $p_{1,2} = \pm j\omega$ & $C_2 = C_1^*$ ; $C_1 = C e^{j\theta}$	Undamped or Oscillatory Unstable

# *Solutions to the Source Free Response of 2<sup>nd</sup> Order ODE*

Overdamped  $y_{sf}(t) = C_1 e^{p_1 t} + C_2 e^{p_2 t}$ ; note that  $p_1$  and  $p_2$  are negative

Critically Damped  $y_{sf}(t) = (C_1 t + C_2) e^{pt}$ ; note that  $p$  is negative

Underdamped  $y_{sf}(t) = C_1 e^{p_1 t} + C_2 e^{p_2 t} = e^{-\alpha t} (C_1 e^{j\omega t} + C_2 e^{-j\omega t})$  where  $p_{1,2} = -\alpha \pm j\omega$

but we can show that

$C_2 = C_1^*$ ; and assuming

$C_1 = a + jb = Ce^{j\theta}$ ; then  $C_2 = a - jb = Ce^{-j\theta}$

$$y_{sf}(t) = e^{-\alpha t} (Ce^{j\theta} e^{j\omega t} + Ce^{-j\theta} e^{-j\omega t}) = e^{-\alpha t} (Ce^{j(\omega t + \theta)} + Ce^{-j(\omega t + \theta)})$$

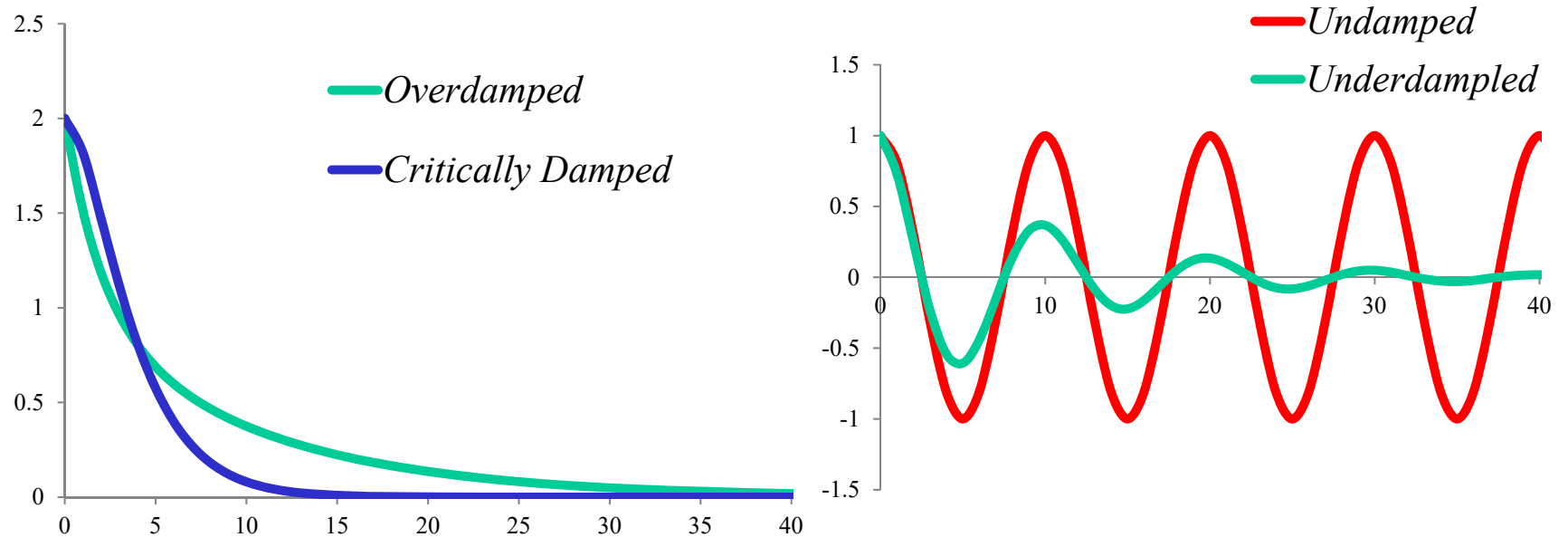
Undamped

$$= e^{-\alpha t} C(e^{j(\omega t + \theta)} + e^{-j(\omega t + \theta)}) = e^{-\alpha t} 2C \cos(\omega t + \theta)$$

$$y_{sf}(t) = C_1 e^{j\omega t} + C_2 e^{-j\omega t} = 2C \cos(\omega t + \theta); \text{ where } p_{1,2} = \pm j\omega; C_2 = C_1^*; C_1 = Ce^{j\theta}$$



# *Source Free Responses of 2<sup>nd</sup> Order ODE*



# *Proof of the Complex Conjugate Constants Underdamped Case*

$$y(t) = e^{-\alpha t} (C_1 e^{j\omega t} + C_2 e^{-j\omega t}) = C_1 e^{s_1 t} + C_2 e^{s_1^* t}; \text{ where } s_1 = -\alpha + j\omega$$

$$y(0) = C_1 + C_2$$

$$\dot{y}(t) = s_1 C_1 e^{s_1 t} + s_1^* C_2 e^{s_1^* t}$$

$$\dot{y}(0) = s_1 C_1 + s_1^* C_2$$

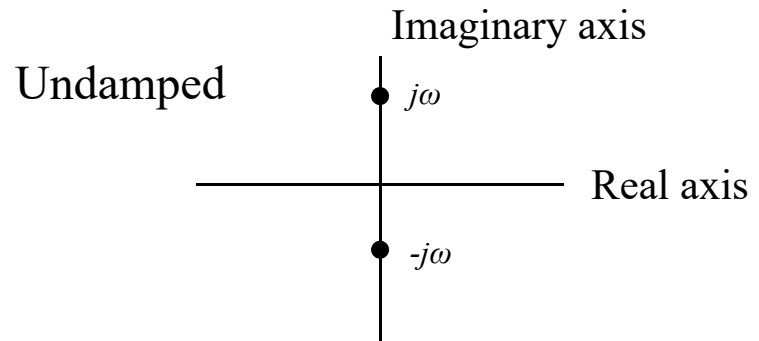
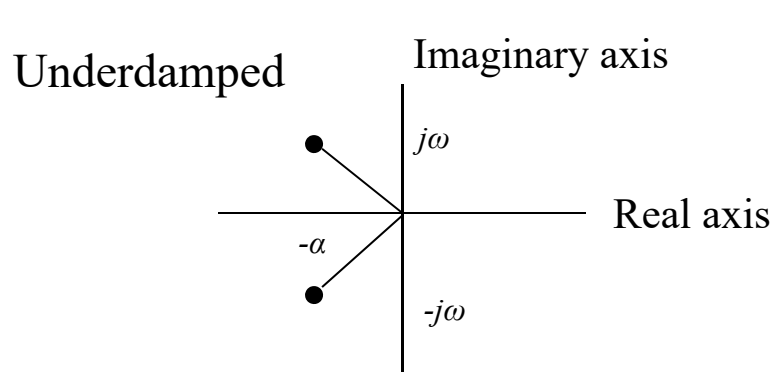
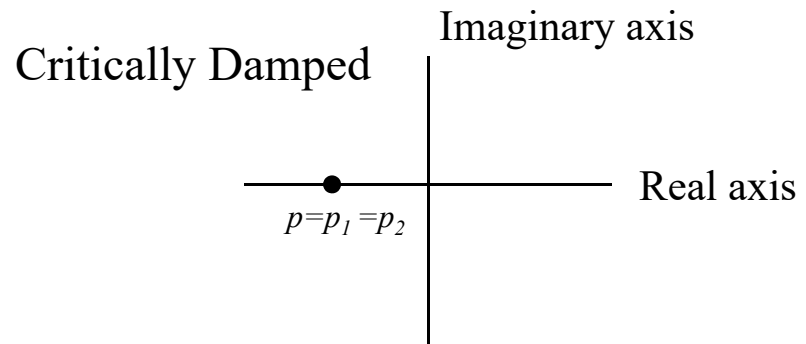
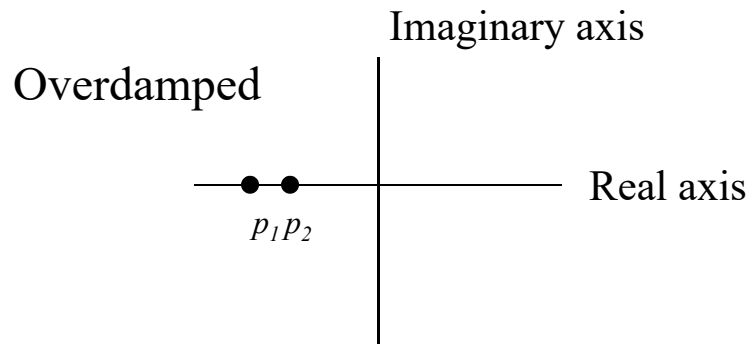
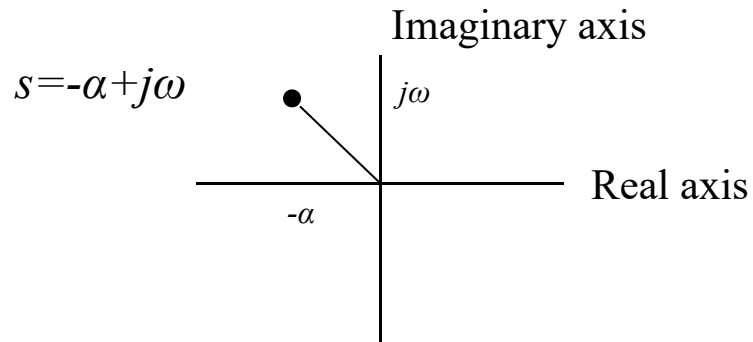
$$C_1 = y(0) - C_2$$

$$\dot{y}(0) = s_1 [y(0) - C_2] + s_1^* C_2$$

$$C_2 = \frac{\dot{y}(0) - s_1 y(0)}{s_1^* - s_1} = \frac{\dot{y}(0) - (-\alpha + j\omega)y(0)}{-j2\omega} = \frac{y(0)}{2} + j \frac{\dot{y}(0) + \alpha y(0)}{2\omega}$$

$$C_1 = y(0) - \frac{\dot{y}(0) - s_1 y(0)}{s_1^* - s_1} = \frac{s_1^* y(0) - \dot{y}(0)}{s_1^* - s_1} = \frac{(-\alpha - j\omega)y(0) - \dot{y}(0)}{-j2\omega} = \frac{y(0)}{2} - j \frac{\dot{y}(0) + \alpha y(0)}{2\omega}$$

# Complex Plane



# *Complete Response of 2<sup>nd</sup> Order ODE*

## *An Example*

••      •      •  
 $y'' + 4y' + 3y = 4e^{-2t}; y(0) = 2; y'(0) = 4$  System with initial conditions

$p^2 + 4p + 3 = 0$  Characteristic Equation

$p_{1,2} = -3, -1$  Eigenvalues - Overdamped

$y_{sf}(t) = A_1 e^{-3t} + A_2 e^{-t}$  Eigenfunctions and source free response

$y_s(t) = A_3 e^{-2t}$  Source free response

$y(t) = A_1 e^{-3t} + A_2 e^{-t} + A_3 e^{-2t}$  Total response

•  
 $y(t) = -3A_1 e^{-3t} - A_2 e^{-t} - 2A_3 e^{-2t}$

••  
 $y(t) = 9A_1 e^{-3t} + A_2 e^{-t} + 4A_3 e^{-2t}$

# *Complete Response of 2<sup>nd</sup> Order ODE Solution*

Substitute the solution, the first and second derivatives into the system

$$(9A_1e^{-3t} + A_2e^{-t} + 4A_3e^{-2t}) + 4(-3A_1e^{-3t} - A_2e^{-t} - 2A_3e^{-2t}) + 3(A_1e^{-3t} + A_2e^{-t} + A_3e^{-2t}) = 4e^{-2t}$$

Resort the equation into like terms

$$(9 - 12 + 3)A_1e^{-3t} + (1 - 4 + 3)A_2e^{-t} + (4 - 8 + 3)A_3e^{-2t} = 4e^{-2t}$$

Notice that the coefficients of the eigenfunctions are zero since they must satisfy the Homogeneous Equation.

What is left is the source term, the response due to the sources:

$$(0)A_1e^{-3t} + (0)A_2e^{-t} - A_3e^{-2t} = 4e^{-2t}$$

From this  $A_3$  is determined:

$$-A_3e^{-2t} = 4e^{-2t} \Rightarrow A_3 = -4$$

Using the initial conditions of the response and the first derivative,  $A_1$  and  $A_2$  are determined.

$$y(0) = A_1 + A_2 - 4 = 2 \Rightarrow A_1 + A_2 = 6$$

$$\dot{y}(0) = -3A_1 - A_2 + 8 = 4 \Rightarrow -3A_1 - A_2 = -4$$

$$-2A_1 = 2 \Rightarrow A_1 = -1; A_2 = 7$$

The total solution is:

$$\therefore y(t) = -e^{-3t} + 7e^{-t} - 4e^{-2t}$$

# *Natural Frequency and Damping Ratio*

$$ap^2 + bp + c = 0$$

$$p^2 + \frac{b}{a}p + \frac{c}{a} = 0$$

$$p_{1,2} = -\frac{b}{2a} \pm \sqrt{\left(\frac{b}{2a}\right)^2 - \frac{c}{a}}$$

$$\text{If } b = 0, \text{ then } p_{1,2} = \pm j\sqrt{\frac{c}{a}} = \pm j\omega_o$$

$$\text{Define } \frac{c}{a} = \omega_o^2$$

$$\text{Define } \frac{b}{2a} = \xi\omega_o \Rightarrow \xi = \frac{b}{2a\omega_o} = \frac{b}{2a\sqrt{\frac{c}{a}}} = \frac{b}{2\sqrt{ac}}$$

We call  $\omega_o$  the undamped natural frequency and  $\xi$  the damping ratio

$$p^2 + 2\xi\omega_o p + \omega_o^2 = 0 \Rightarrow p_{1,2} = -\frac{b}{2a} \pm \sqrt{\left(\frac{b}{2a}\right)^2 - \frac{c}{a}} \Rightarrow -\xi\omega_o \pm \sqrt{(\xi\omega_o)^2 - \omega_o^2}$$

$$p_{1,2} = -\xi\omega_o \pm \sqrt{(\xi\omega_o)^2 - \omega_o^2} = -\xi\omega_o \pm \omega_o\sqrt{\xi^2 - 1}$$

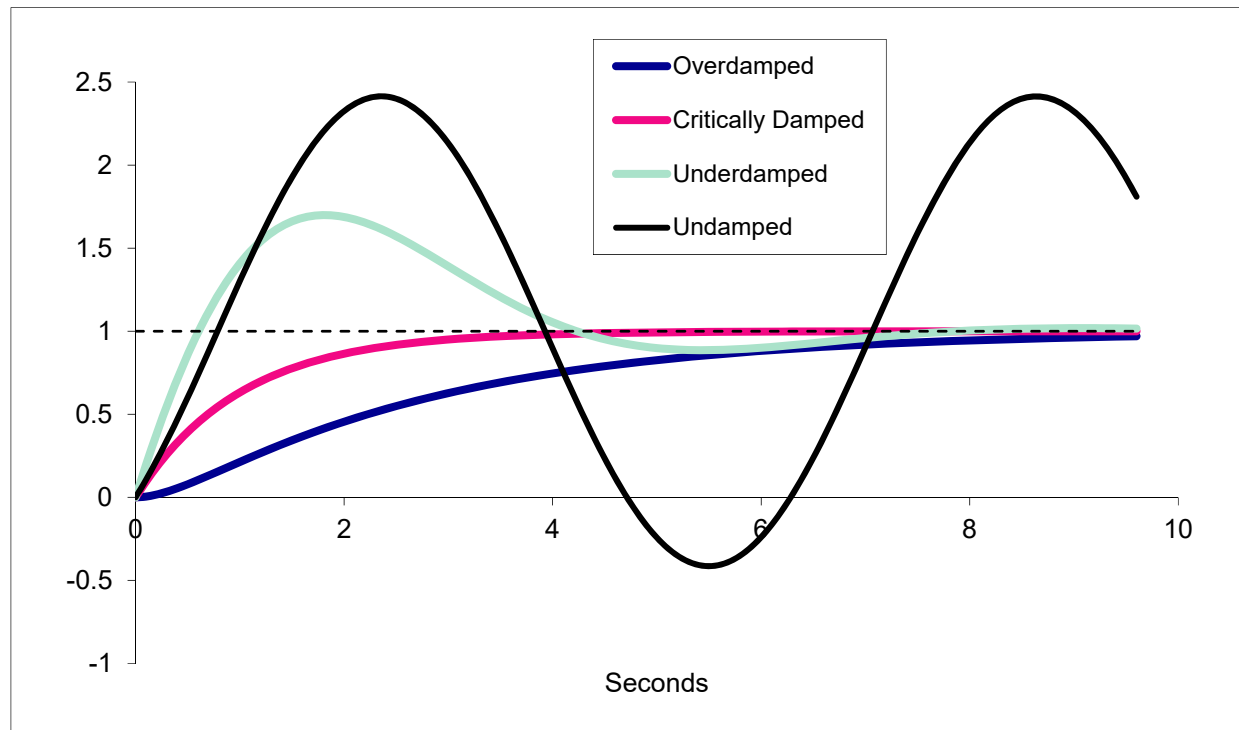
$$\text{For } \xi > 1, \text{ overdamped, } p_{1,2} = -\xi\omega_o \pm \omega_o\sqrt{\xi^2 - 1} = \omega_o(-\xi \pm \sqrt{\xi^2 - 1})$$

$$\text{For } \xi = 1, \text{ critically damped, } p_{1,2} = -\omega_o$$

$$\text{For } \xi < 1, \text{ underdamped, } p_{1,2} = -\xi\omega_o \pm j\omega_o\sqrt{1 - \xi^2}$$

$$\text{For } \xi = 0, \text{ undamped, } p_{1,2} = \pm j\omega_o$$

# *Natural Frequency and Damping Ratio Unit Step Response*



# Unit Step Response

## Overdamped

$$p^2 y(t) + \frac{b}{a} p y(t) + \frac{c}{a} y(t) = \frac{1}{a} x(t)$$

$$p^2 y(t) + 2\xi\omega_o p y(t) + \omega_o^2 y(t) = Kx(t) = 1$$

$$\text{Source response is } y_s(t) = A = \frac{1}{\omega_o^2}$$

$$\text{For } \xi > 1, \text{ overdamped, } p_{1,2} = (-\xi \pm \sqrt{\xi^2 - 1})\omega_o$$

$$\text{Source free response is } y_{sf}(t) = C_1 e^{(-\xi - \sqrt{\xi^2 - 1})\omega_o t} + C_2 e^{(-\xi + \sqrt{\xi^2 - 1})\omega_o t}$$

$$y(t) = \frac{1}{\omega_o^2} + C_1 e^{(-\xi - \sqrt{\xi^2 - 1})\omega_o t} + C_2 e^{(-\xi + \sqrt{\xi^2 - 1})\omega_o t}$$

$$y(0) = 0 = \frac{1}{\omega_o^2} + C_1 + C_2; C_1 = -\left(\frac{1}{\omega_o^2} + C_2\right)$$

$$\dot{y}(0) = 0 + (-\xi - \sqrt{\xi^2 - 1})\omega_o C_1 + (-\xi + \sqrt{\xi^2 - 1})\omega_o C_2$$

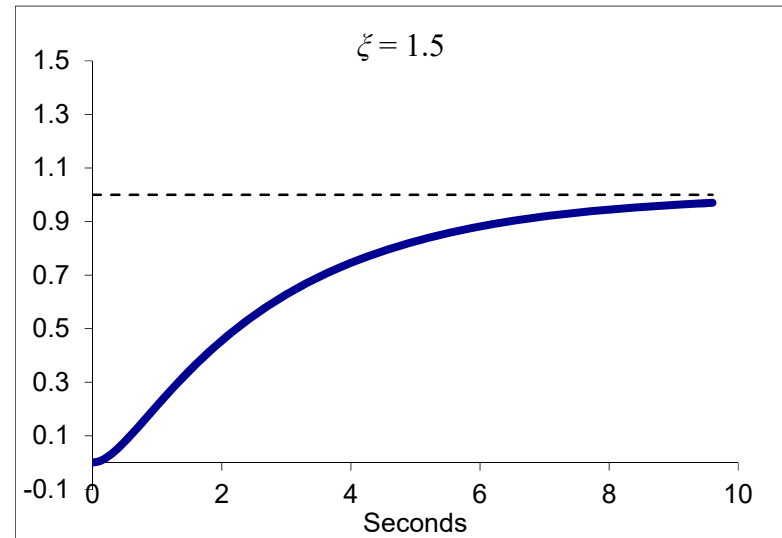
$$\dot{y}(0) = -(-\xi - \sqrt{\xi^2 - 1})\omega_o \left[\left(\frac{1}{\omega_o^2} + C_2\right)\right] + (-\xi + \sqrt{\xi^2 - 1})\omega_o C_2 = (\xi + \sqrt{\xi^2 - 1})\omega_o \left[\left(\frac{1}{\omega_o^2} + C_2\right)\right] + (-\xi + \sqrt{\xi^2 - 1})\omega_o C_2$$

$$= -(-\xi - \sqrt{\xi^2 - 1})\omega_o \frac{1}{\omega_o^2} + (\xi + \sqrt{\xi^2 - 1})\omega_o C_2 + (-\xi + \sqrt{\xi^2 - 1})\omega_o C_2 = (\xi + \sqrt{\xi^2 - 1})\frac{1}{\omega_o} + (2\sqrt{\xi^2 - 1})\omega_o C_2$$

$$\dot{y}(0) = -(-\xi - \sqrt{\xi^2 - 1})\frac{1}{\omega_o} + (2\sqrt{\xi^2 - 1})\omega_o C_2 \Rightarrow C_2 = \frac{\dot{y}(0) + (-\xi - \sqrt{\xi^2 - 1})\frac{1}{\omega_o}}{(2\sqrt{\xi^2 - 1})\omega_o}$$

$$C_2 = \frac{\dot{y}(0)\omega_o + (-\xi - \sqrt{\xi^2 - 1})}{2\omega_o^2 \sqrt{\xi^2 - 1}}; C_1 = \left(\frac{\xi - \sqrt{\xi^2 - 1} - \dot{y}(0)\omega_o}{2\omega_o^2 \sqrt{\xi^2 - 1}}\right)$$

$$y(t) = \frac{1}{\omega_o^2} + \frac{-\dot{y}(0)\omega_o + \xi - \sqrt{\xi^2 - 1}}{2\omega_o^2 \sqrt{\xi^2 - 1}} e^{(-\xi - \sqrt{\xi^2 - 1})\omega_o t} + \frac{\dot{y}(0)\omega_o - \xi - \sqrt{\xi^2 - 1}}{2\omega_o^2 \sqrt{\xi^2 - 1}} e^{(-\xi + \sqrt{\xi^2 - 1})\omega_o t}$$





# Unit Step Response Critically Damped

$$p^2 y(t) + \frac{b}{a} p y(t) + \frac{c}{a} y(t) = \frac{1}{a} x(t)$$

$$p^2 y(t) + 2\xi\omega_o p y(t) + \omega_o^2 y(t) = Kx(t) = 1$$

Source response is  $y_s(t) = A = \frac{1}{\omega_o^2}$

For  $\xi = 1$ , critically damped,  $p_{1,2} = -\omega_o$

Source free response is  $y_{sf}(t) = (C_1 t + C_2) e^{-\omega_o t}$

$$y(t) = \frac{1}{\omega_o^2} + (C_1 t + C_2) e^{-\omega_o t}$$

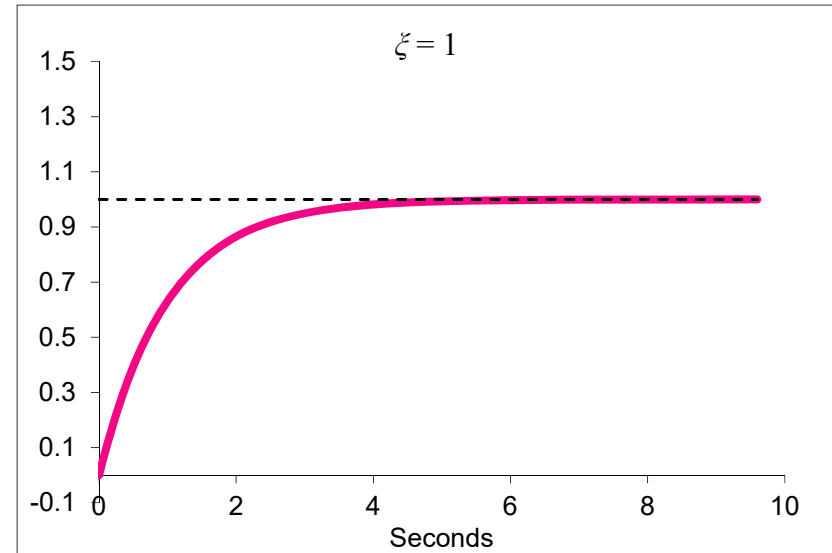
$$y(0) = 0 = \frac{1}{\omega_o^2} + C_2$$

$$C_2 = -\frac{1}{\omega_o^2}$$

$$\dot{y}(0) = 0 + C_1 - \omega_o C_2$$

$$C_1 = \dot{y}(0) + \omega_o C_2 = \dot{y}(0) - \frac{1}{\omega_o}$$

$$y(t) = \frac{1}{\omega_o^2} + \left( \left[ \dot{y}(0) - \frac{1}{\omega_o} \right] t - \frac{1}{\omega_o^2} \right) e^{-\omega_o t}$$



# Unit Step Response

## Underdamped

$$p^2 y(t) + \frac{b}{a} p y(t) + \frac{c}{a} y(t) = \frac{1}{a} x(t)$$

$$p^2 y(t) + 2\xi\omega_o p y(t) + \omega_o^2 y(t) = Kx(t) = 1$$

Source response is  $y_s(t) = A = \frac{1}{\omega_o^2}$

For  $\xi < 1$ , underdamped,  $p_{1,2} = -\xi\omega_o \pm j\omega_o\sqrt{1-\xi^2}$

Source free response is  $y_{sf}(t) = e^{-\xi\omega_o t} K \cos(\omega_o\sqrt{1-\xi^2}t + \theta)$

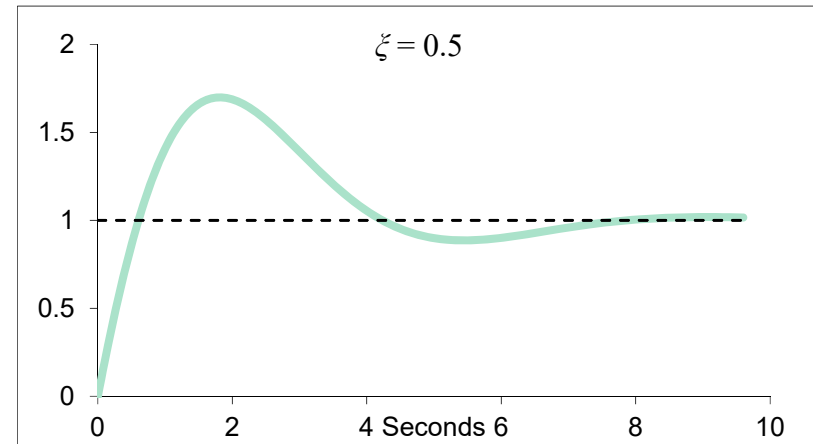
$$y(t) = \frac{1}{\omega_o^2} + e^{-\xi\omega_o t} K \cos(\omega_o\sqrt{1-\xi^2}t + \theta)$$

$$y(0) = 0 = \frac{1}{\omega_o^2} + K \cos(\theta); K \cos(\theta) = -\frac{1}{\omega_o^2}$$

$$\dot{y}(0) = 0 + -\xi\omega_o K \cos(\theta) - \omega_o\sqrt{1-\xi^2} K \sin(\theta) = \xi\omega_o \frac{1}{\omega_o^2} - \omega_o\sqrt{1-\xi^2} K \sin(\theta); K \sin(\theta) = -\frac{\dot{y}(0) - \frac{\xi}{\omega_o}}{\omega_o\sqrt{1-\xi^2}}$$

$$\theta = \tan^{-1}\left(\frac{\dot{y}(0)\omega_o - \xi}{\sqrt{1-\xi^2}}\right); K = -\frac{1}{\omega_o^2 \cos(\tan^{-1}\left(\frac{\dot{y}(0)\omega_o - \xi}{\sqrt{1-\xi^2}}\right))}$$

$$y(t) = \frac{1}{\omega_o^2} - e^{-\xi\omega_o t} \frac{1}{\omega_o^2 \cos(\tan^{-1}\left(\frac{\dot{y}(0)\omega_o - \xi}{\sqrt{1-\xi^2}}\right))} \cos(\omega_o\sqrt{1-\xi^2}t + \tan^{-1}\left(\frac{\dot{y}(0)\omega_o - \xi}{\sqrt{1-\xi^2}}\right))$$



# Unit Step Response

## Undamped

$$p^2 y(t) + \frac{b}{a} p y(t) + \frac{c}{a} y(t) = \frac{1}{a} x(t)$$

$$p^2 y(t) + 2\xi\omega_o p y(t) + \omega_o^2 y(t) = Kx(t) = 1$$

Source response is  $y_s(t) = A = \frac{1}{\omega_o^2}$

For  $\xi = 0$ , undamped,  $p_{1,2} = \pm j\omega_o$

Source free response is  $y_{sf}(t) = C \cos(\omega_o t + \theta)$

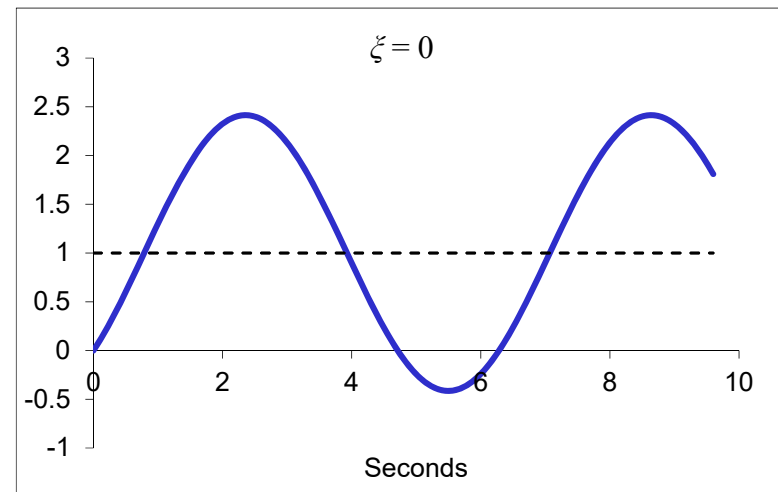
$$y(t) = \frac{1}{\omega_o^2} + C \cos(\omega_o t + \theta)$$

$$y(0) = 0 = \frac{1}{\omega_o^2} + C \cos(\theta); C \cos(\theta) = -\frac{1}{\omega_o^2}$$

$$\dot{y}(0) = 0 = -\omega_o C \sin(\theta); C \sin(\theta) = -\frac{\dot{y}(0)}{\omega_o}$$

$$\theta = \tan^{-1}(\dot{y}(0)\omega_o); C = -\frac{1}{\omega_o^2 \cos(\tan^{-1}(\dot{y}(0)\omega_o))}$$

$$y(t) = \frac{1}{\omega_o^2} - \frac{1}{\omega_o^2 \cos(\tan^{-1}(\dot{y}(0)\omega_o))} \cos(\omega_o t + \tan^{-1}(\dot{y}(0)\omega_o))$$



# *Poles and Zeroes*

- Source Response:

$$A(p)y(t) = B(p)x(t)$$

$$y(t) = \frac{B(p)}{A(p)}x(t)$$

$$y(t) = H(p)x(t)$$

$$H(p) = \frac{B(p)}{A(p)}$$

*H(p) is known as the system function or network response*

- We can think of the solutions of  $B(p) = 0$  as the zeroes of the system
- We can think of the solutions of  $A(p) = 0$  as the poles of the system

## Poles and Zeros Continued

- If we assume that the source  $x(t)$  has the form  $x(t)=e^{st}$  where  $s$  is a complex number, we can plot the poles and zeroes in the complex plane to graphically see the response to a particular source function. Note that  $H(s)$  is a complex number with magnitude and angle.

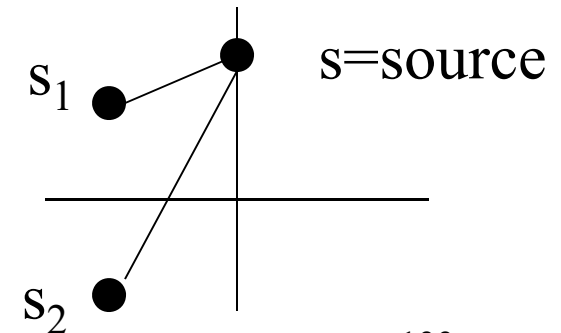
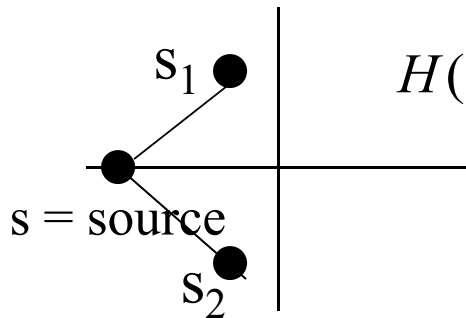
$$y(t) = H(p)x(t) = H(p)e^{st}$$

$$= H(s)e^{st}$$

$$\text{since } pe^{st} = \frac{d}{dt}e^{st} = se^{st}$$

For example, if  $H(p)$  have no zeroes and 2 poles, then

$$H(s) = \frac{A}{(s - s_1)(s - s_2)}$$



# Homework

- 2<sup>st</sup> Order ODEs

1. Using Matlab, plot the response for following systems. Identity what type of system each is. Submit your code:

a)  $\ddot{x} + 10\dot{x} + 4x = 0; x(0) = 5; \dot{x}(0) = 0$

b)  $\ddot{x} + 4\dot{x} + 4x = 0; x(0) = 5; \dot{x}(0) = 0$

c)  $\ddot{x} + 1\dot{x} + 4x = 0; x(0) = 5; \dot{x}(0) = 0$

d)  $\ddot{x} + 4x = 0; x(0) = 5; \dot{x}(0) = 0$

2. 2.2 An LTI system is described by the second-order ODE:

$$\ddot{y} + 7\dot{y} + 10y = x(t)$$

- a. Use the p operator notation to find the roots of the characteristic equation.
- b. Assume  $y(0)=0$ ,  $dy(0)/dt=9$  and  $x(t)=0$ , find  $y(t)$ .
- c. Now Let  $x(t) = 10$  and same initial conditions, find  $y(t)$ .

## *Homework*

3. A quadratic low-pass filter is described by the second-order ODE:

$$\ddot{y} + (2\xi\omega_n)\dot{y} + \omega_n^2 y = x(t)$$

- a. The characteristic equation for this ODE in terms of the  $p$  operator has complex conjugate roots. Find an algebraic expression for the position of the roots.
- b. Let  $x(t)=1$ . Find the steady-state output.
- c. Let  $x(t)=0$ ,  $y(0)=0$  and  $dy(0)/dt=10$ , Find and sketch  $y(t)$  for  $\xi =.5$  and  $\omega_n=1$

## *Homework*

4. An alternate way of writing the ODE for an underdamped system is

$$\ddot{y} + 2a\dot{y} + (b^2 + a^2)y = x(t)$$

Let  $a=.5$  and  $b=.86603$ . Repeat a, b, and c in problem 2.



# *Homework*

## 5. BioSignals

- An heart signal is sampled at the rate 250 s/s and is passed to EKG which has an input consisting of a low pass filter. The filter is a resistor and capacitor in series where the output of the filter is taken across the capacitor. What should be the value of the Capacitor if the Resistor is 1k ohms and the time constant of the filter so that the transient response is completed within 1/10 of the sample time? What is the cutoff frequency of this filter?

## *Homework*

6. Respiration may be modeled with the following second order equation, where  $y(t)$  is the respiration signal and  $x(t)$  is the additional load above the resting respiration on the body:

$$\ddot{y} + \omega_n^2 y = x(t)$$

- a. Calculate the resting respiration rate. Assume  $y(0)=0$  and  $dy(0)/dt=\omega_n$ .
- b. Calculate the respiration rate when  $x(t)=\cos(\omega_n t)$ . Assume the same initial conditions as part a.
- c. Use Matlab to graph the signals for both parts. Assume  $\omega_n=2\pi$ .